

**STK4030V**

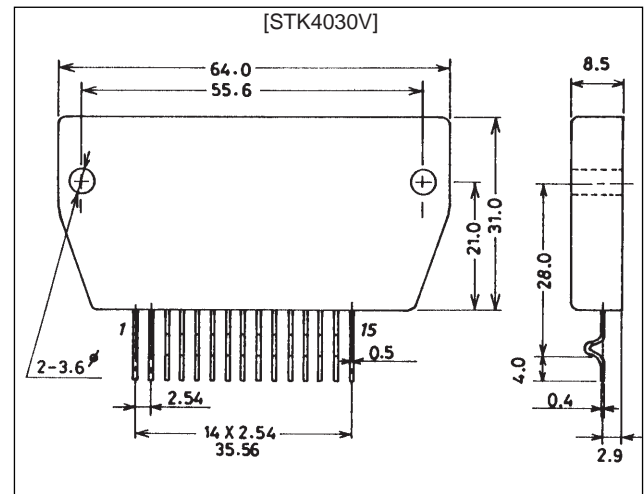
AF Power Amplifier (Split Power Supply) (35 W min, THD = 0.08%)

Features

- Compact packaging supports slimmer set designs (up to 70 W)
- Series designed for 20 up to 100 W (200 W) and pin-compatibility (120 to 200 W have 18 pins)
- Simpler heat sink design facilitates thermal design of slim stereo sets
- Current mirror circuit application reduces distortion to 0.08%
- Supports addition of electronic circuits for thermal shutdown and load-short protection circuit as well as pop noise muting which occurs when the power supply switch is turned on and off

Package Dimensions

unit: mm

4062

Specifications

Maximum Ratings at $T_a = 25^\circ\text{C}$

Parameter	Symbol	Conditions	Ratings	Unit
Maximum supply voltage	$V_{CC \text{ max}}$		± 45	V
Thermal resistance	θ_{j-c}		2.1	$^\circ\text{C/W}$
Junction temperature	T_j		150	$^\circ\text{C}$
Operating substrate temperature	T_c		125	$^\circ\text{C}$
Storage temperature	T_{stg}		-30 to $+125$	$^\circ\text{C}$
Available time for load shorted	t_s^*	$V_{CC} = \pm 30 \text{ V}$, $R_L = 8 \ \Omega$, $f = 50 \text{ Hz}$, $P_O = 35 \text{ W}$	2	s

Note: Use a constant-voltage power supply as the test power supply unless otherwise specified.

* Use the transformer power supply shown on the next page when measuring the available time for load shorted and the output noise voltage.

Recommended Operating Conditions at $T_a = 25^\circ\text{C}$

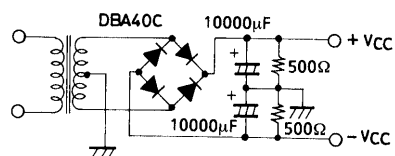
Parameter	Symbol	Conditions	Ratings	Unit
Recommended supply voltage	V_{CC}		± 30	V
Load resistance	R_L		8	Ω

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Operating Characteristics at $T_a = 25^\circ\text{C}$, $V_{CC} = \pm 30\text{ V}$, $R_L = 8\ \Omega$, $V_G = 40\text{ dB}$, $R_g = 600\ \Omega$, 100 k LPF on , R_L (non-inductive)

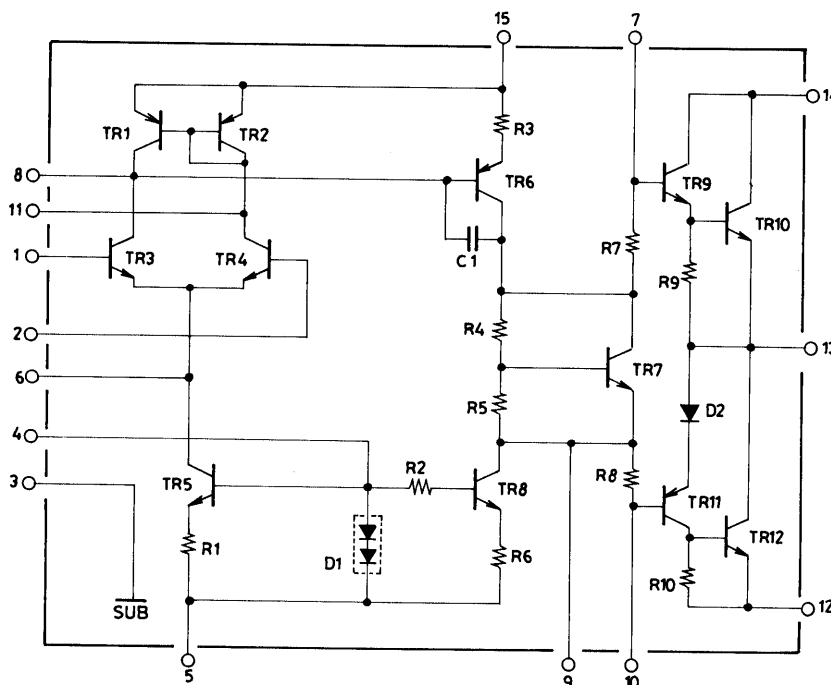
Parameter	Symbol	Conditions	Ratings			Unit
			min	typ	max	
Quiescent current	I_{CCO}	$V_{CC} = \pm 36\text{ V}$	15		120	mA
Output power	$P_O(1)$	THD = 0.08%, $f = 20\text{ Hz to } 20\text{ kHz}$	35			W
	$P_O(2)$	$V_{CC} = \pm 27\text{ V}$, THD = 0.2%, $R_L = 4\ \Omega$, $f = 1\text{ kHz}$	40			
Total harmonic distortion	THD	$P_O = 1.0\text{ W}$, $f = 1\text{ kHz}$			0.08	%
Frequency response	f_L, f_H	$P_O = 1.0\text{ W}$, -3 dB		20 to 50 k		Hz
Input resistance	r_i	$P_O = 1.0\text{ W}$, $f = 1\text{ kHz}$		55		k Ω
Output noise voltage	V_{NO}^*	$V_{CC} = \pm 36\text{ V}$, $R_g = 10\text{ k}\Omega$		1.2		mVrms
Neutral voltage	V_N	$V_{CC} = \pm 36\text{ V}$	-70	0	+70	mV

Equivalent Circuit



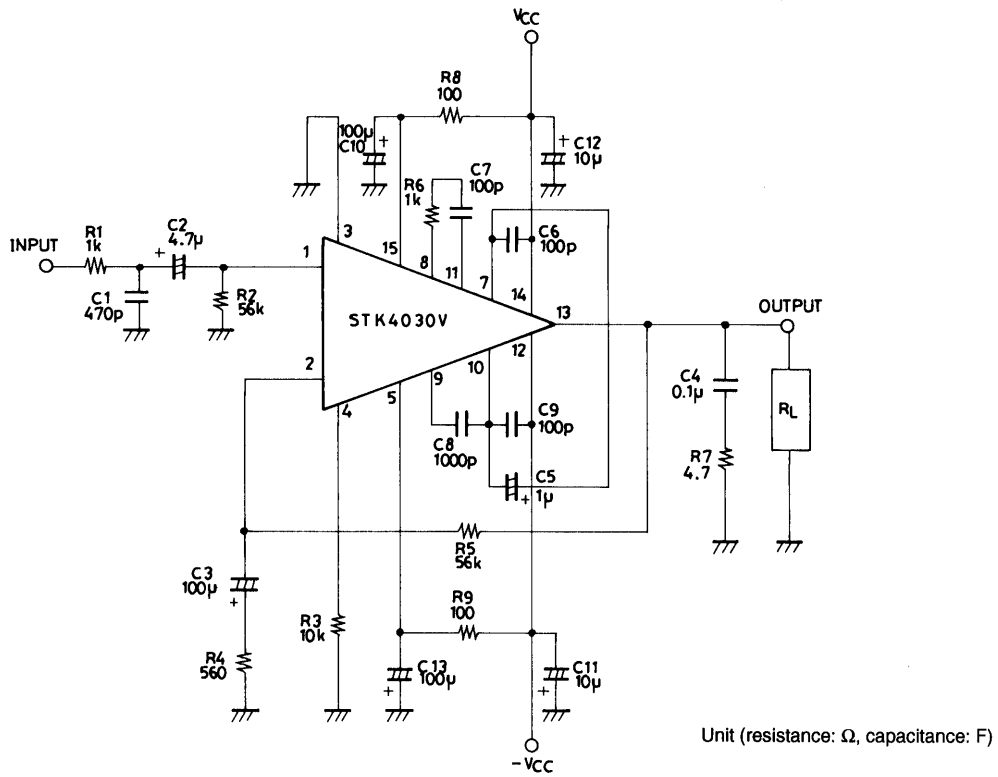
**Specified Transformer Power Supply
(RP-25 Equivalent)**

Equivalent Circuit

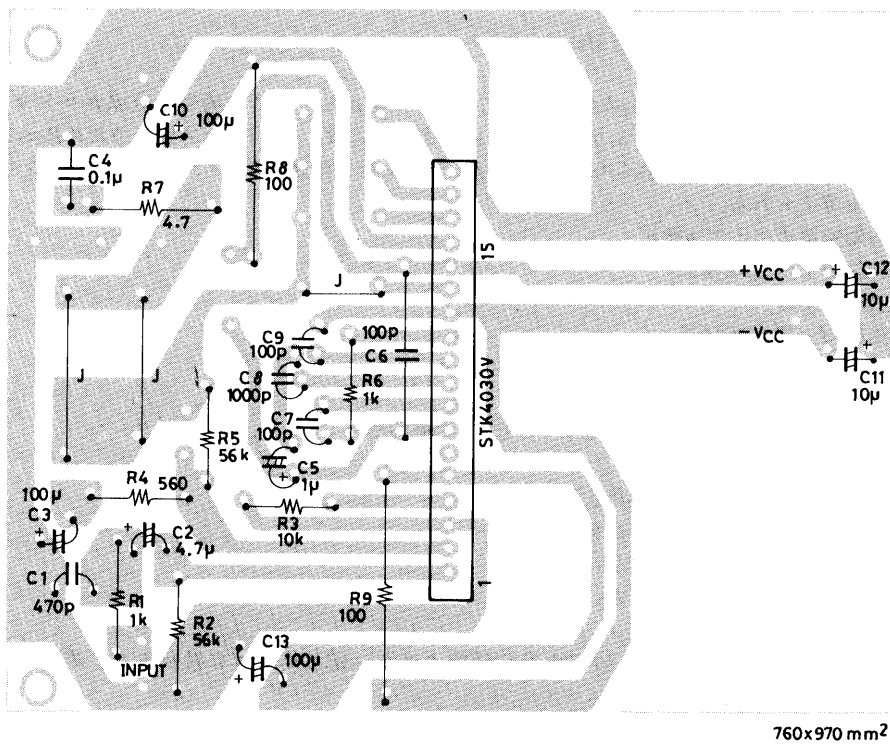


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Application Circuit: 35W min Single Channel AF Power Amplifier



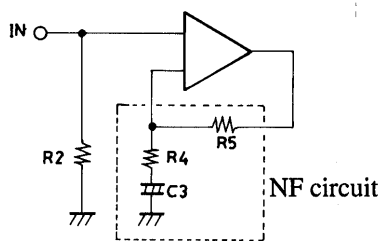
Sample Printed Circuit Pattern for Application Circuit (Copper-foiled side)



Unit (resistance: Ω , capacitance: F)

Description of External Parts

- R₁, C₁ : Input filter circuit
 - Reduces high-frequency noise.
- C₂ : Input coupling capacitor
 - DC current suppression. A reduction in reactance is effective because of increases in capacitor reactance at low frequencies and 1/f noise dependence on signal source resistance which result in output noise worsening.
- R₂ : Input bias resistor
 - Biases the input pin to zero.
 - Effects V_N stability (refer to NF circuit).
 - Due to differential input, input resistance is more or less determined by this resistance value.
- R₄, R₅ : NFB circuit (AC NF circuit). Use of resistor with 1% error is suggested.
- C₃ (R₂)



- C₃ : AC NF capacitor
- R₄, R₅ : Used for VG setting.

- VG settings are obtained using R₄ and R₅ according to the following equation:

$$\log_{20} \cdot \frac{R_5}{R_4} \quad 40 \text{ dB is recommended.}$$

- Low-frequency cutoff frequency settings are obtained using R₄ and C₃ according to the following equation:

$$f_L = \frac{1}{2\pi \cdot R_4 \cdot C_3} \quad [\text{Hz}]$$

When changing the VG setting, you should change R₄ which requires a recheck of the low cutoff frequency setting. When the VG setting is changed using R₅, the setting should ensure R₂ equals R₅ so that V_N balance stability is maintained. If the resistor value is increased more than the existing value, V_N balance may be disturbed and result in deterioration of V_N temperature characteristics.

- R₃ : Differential constant-current bias resistor
- R₆, R₇ : For oscillation suppression and phase compensation applications
(For use with differential stage applications)
- R₇, C₄ : For oscillation suppression and phase compensation applications
(A Mylar capacitor is recommended for C₄ for use with output stage applications)
- C₆, C₉ : For oscillation suppression and phase compensation applications
Power stage (Must be connected near the pin) C₆: Positive (+) power C₉: Negative (-) power
- C₈ : For oscillation suppression and phase compensation applications
(Oscillation suppression before power step clip)
- C₅ : For oscillation suppression and distortion improvement applications
- R₈, C₁₀ : Ripple filter circuit on positive (+) side.
- R₉, C₁₃ : Ripple filter circuit on negative (-) side.
- C₁₁, C₁₂ : For oscillation suppression applications
 - Used for reducing power supply impedance to stable IC operation and should be connected near the IC pin. We recommend that you use an electrolytic capacitor.

This datasheet has been download from:

www.datasheetcatalog.com

Datasheets for electronics components.